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# **Experimental Demonstrations of High-Accuracy** 3D/2D Indoor Visible Light Positioning Using Imaging MIMO Receivers and Artificial Neural Networks

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**Abstract:** This paper proposes and presents the first experimental demonstration of a high-precision indoor 2D and 3D visible light positioning (VLP) system using an imaging multiple-input multiple-output (MIMO) configuration with supervised artificial neural network (ANN). The proposed system utilizes four distributed transmitters and receivers with four photodiodes and an imaging optics. The experiments are conducted in a typical indoor environment with transmitter separations of 300 mm and a link distance of 1400 mm. The experimental results show 2D and 3D positioning accuracies of 3.7 mm and 51 mm, respectively. A simulation model is also developed for the VLP system to verify the experimental results. Further optimization of the VLP system in the simulation platform leads to improved 2D and 3D positioning accuracies of 2 mm and 9.3 mm, respectively. The proposed system can seamlessly converge with existing lighting infrastructures and is also compatible with the imaging MIMO visible light communication (VLC) system, indicating the potential for practical implementation in integrated communications and positioning applications.

#### 1. Introduction

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Over the past decade, visible light communication 31 (VLC) has emerged as a complementary technology to 32 (RF)-based wireless 33 radio frequency communications. VLC leverages existing solid-state lighting 34 infrastructure for high-speed wireless communication, thus 35 offering advantages such as unlicensed spectrum operation, 36 low power consumption, and reduced implementation costs 37 [1]. Operating in the unlicensed spectrum while still 38 maintaining low power consumption, this technology is 39 expected to play a key role in upcoming 6G networks, 40 demonstrating high-speed transmission capabilities and 41 compatibility with various communication systems. In 42 addition, it also offers an alternative for integrating aerial, 43 submarine, and indoor networks for 6G and beyond [2]. 44 Among its many applications, indoor and outdoor visible 45 light positioning and navigation are promising areas.

Positioning technologies have recently attracted 47 significant attention due to their applications in wide-ranging 48 monitoring, surveillance, or tracking. Compared to RF 49 positioning technologies, visible light positioning (VLP) 50 offers unique advantages of high-accuracy due to shorter 51 wavelengths and less sensitivity to multipath propagation, 52 zero electromagnetic interference (EMI), and dual 53 functionalities of illumination and positioning [1].

VLP systems utilize photodiodes (PDs) or image 55 sensors (IS) receivers to provide indoor positioning solutions. 56 PD-based VLP systems offer advantages such as fast

response times, compatibility with communications, and suitability for real-time applications and varying lighting conditions. Existing PD-based VLP techniques can be classified into two main categories: distance-based and distance-free [3], [4]. Distance-based techniques include the use of received signal strength (RSS) to estimate the distance between the receiver (Rx) and the transmitter (Tx) [5]. Other distance-based techniques employ time of arrival (TOA) [6] and time difference of arrival (TDOA) [7]; both require precise synchronization between transmitters and receivers, resulting in increased system complexity. Furthermore, angle of arrival (AOA) based techniques require diversified angles and a relatively large number of receiver devices to operate effectively [8]. Distance-free techniques are independent of geometric distance measurements, but often require more complex hardware and configuration. One commonly adopted distance-free technique uses RSS values as fingerprint features for indoor positioning [9].

In contrast to PD-based VLP systems, IS-based VLP systems capture the images of the modulated intensity of LED luminaires and subsequently process them using image processing algorithms to estimate the position [10]. This enables the extraction of detailed features, the improvement in interference rejection, and the elimination of multipath reflections [11]. Furthermore, cameras are extensively incorporated into consumer devices, such as smartphones, enabling VLP systems to use existing hardware. However, these systems are subject to several challenges, including

CONTINUATION	i proposed work.							
Reference/year	Technique	Type of Study	Number of Txs	Number of PDs	Room dimensions (mm) L×W×H	3D	Compatible with communications	Accuracy
[17], 2018	RSS	Simulation	1	4	3000×3000×1250	No	No	35.0 mm @ 90% CDF
[18], 2021	RSS fingerprinting with fabricated data and ML		4	4	1200×1200×1600	No	No	8.3 mm @ 90% CDF
[19], 2022	AOA with LSTMNN	Experimental	1	4	400×400	No	No	29.0 mm @90% CDF
[20], 2018	Three different ANNs (one for each axis)	Simulation	16	361 (19×19)	4000×4000×3000	Yes	No	0.4 mm @90% CDF
[21], 2022	RSS fingerprinting (ML) ANN	Simulation	4	1-4	5000×5000×5000	Yes	No	19.8 mm @ 90% CDF (LOS-3D) 10.3 mm @ 90% CDF (LOS-2D)
[22], 2024	Deep residual shrinkage network (DRSN)	Experimental & Simulation		4	3600×3600×3000	Yes	No	23.5 mm @90% CDF (Simulation) 100 mm @90% CDF (Experimental)
Our work	RSS fingerprinting MLP-ANN with Imaging receiver	Experimental & Simulation	4	4	400×400×120	Yes	Yes	2 mm (2D) @90% CDF 9 mm (3D) @90% CDF

slower response times generally unsuitable for high-speed 39 communication due to their limited frame rates and 40 sensitivity to lighting conditions, which can affect the 41 accuracy and difficulty of employing them in real-time 42 systems, which limits their effectiveness in specific scenarios. 43

Advanced solutions for accurate VLP have been 44 proposed based on machine learning and deep learning 45 (ML/DL), i.e., linear or higher-order regression [12]. Various 46 algorithms, such as K-nearest neighbor (KNN), support 47 vector machine (SVM), and artificial neural network (ANN) 48 [13], are showcasing promising outcomes and achieving mm 49 levels of accuracy, thereby offering potential solutions for 50 VLP. These techniques are often used in fingerprint-based 51 systems, where a database of RSS values and their 52 corresponding coordinates is pre-collected. ANNs, for 53 instance, are trained using the offline fingerprint data. Once 54 trained, ANNs can accurately predict a user's location in real-55 time based on new RSS measurements. The Multi-Layer 56 Perceptron (MLP) network, often configured with a single 57 hidden layer, is a commonly chosen ANN architecture for 58 conducting localization tasks [14], [15].

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The spatial diversity provided by multiple PDs 60 improves the robustness of the positioning system against 61 obstacles and interference, which are common factors in 62 indoor environments [16]. Moreover, the literature shows the 63 use of multiple PDs in experimental VLP systems to improve 64 positioning accuracy, as detailed in Table I. For example, 65 tilted PDs were proposed to improve the accuracy of VLP 66 systems [17], where a localization error of 35 mm was 67 obtained. Furthermore, a ML technique is combined with 68 multiple detectors to provide higher accuracy. In [18], four 69 PDs and RSS-based fingerprinting with a Weighted K-nearest 70 neighbors (WkNN) algorithm were employed to demonstrate 71 positioning errors of 8.3 mm and 20.45 mm with four and two 72 luminaires, respectively. A VLP system based on a single 73 LED and multiple silicon solar cells employing AOA and a 74 long short-term memory neural network (LSTMNN) model 75 has achieved an average positioning error of 17.8 mm, and 90%6

of the experimental data had a positioning error within ~29 mm [19]. A theoretical approach based on 16 LED lamps and a grid of 361 receivers with three ANNs to estimate 3D positioning from RSS has achieved an average distance error of 0.4 mm [20]. In [21], the authors proposed a VLP system based on four evenly distributed LED emitters and a MLP for 2D positioning with an estimation of the root-mean-square (RMS) errors as 10.3 mm and 13.3 mm for LOS and non-LOS links, and 19.8 mm and 21 mm for 3D localization. In [22], the authors propose a deep residual shrinkage network (DRSN) with a single LED and 4-PDs. The system achieves an accuracy where 90% of the errors are below 23.5 mm in simulations and below 100 mm in experiments.

The LiFi-based integrated communication and positioning paradigm is expected to be a key technology for 6G networks [23]. In VLC/LiFi technologies, numerous studies have shown high-speed communications utilizing multiple-input multiple-output (MIMO) configurations and accurate VLP positioning facilitated by a distributed illumination infrastructure [24]. The optical MIMO receiver be realized using imaging and non-imaging configurations [25]. Imaging MIMO systems are preferred over non-imaging MIMO configurations due to enhanced data rate scalability, compactness, and a well-conditioned channel H-matrix [25], [26]. However, to the best of the authors' knowledge, there has been no previously published work demonstrating imaging MIMO configurations for highly accurate VLP. This paper is the first attempt to showcase highly accurate 2D/3D VLP using imaging MIMO setups. While this paper primarily focuses on proof-ofconcept demonstrations of VLP using such configurations, the overarching objective is to exhibit integrated communications and positioning in future iterations.

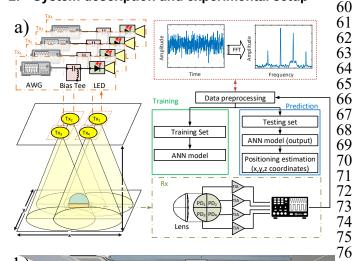
Hence, the novelty and original contributions of this paper are as follows:

 To the best of the authors' knowledge, this is the first experimental demonstration and simulation study of imaging optical MIMO configuration for 2D/3D

- positioning. Furthermore, this is the first study of 33 utilizing the supervised ANN with imaging MIMO for 34
- This paper provides a comprehensive study of the impact 36 of defocusing the lens position in the receiver on VLP 37 performance; clearly demonstrating the trade-off 38 between the field of view (FOV) and positioning 39 accuracy
- 23456789 This is the first experimental demonstration to verify the 41 10 improved 2D and 3D positioning accuracy due to 42 11 imaging spatial diversity provided by multiple PDs. 12
  - This work demonstrates the potential of imaging MIMO 44 VLP configuration for integrated sensing and 45 communication applications, aligning with future 6G 46 network requirements.

The subsequent sections of this work are organized as 48follows: section 2 provides a detailed description of the 49 the 50 VLC positioning system, including experimental setup and the signal processing for the ANN 51 algorithm. Section 3 presents the results obtained from both 52 laboratory measurements and simulations evaluating the 53 accuracy of the 2D/3D positioning system. Finally, section 454 presents the conclusions of this work by summarizing the 55 feasibility and accuracy of the proposed 2D/3D VLP system 56 57 and identifying the main challenges for further research. 58

#### 26 System description and experimental setup



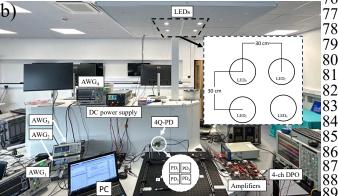


Fig. 1. Multi-PD VLP system with an imaging receiver: (a) Schematic diagram; (b) Photograph of the laboratory setup. The insets show the geometrical distribution of the transmitters and receivers.

Fig. 1(a) depicts the experimental setup of the VLP system proposed in this work, including the Tx and Rx configurations, the signal processing procedure at the Rx, and position estimation based on ANN, which will be described in detail in this section. The experimental parameters are summarized in Table II. The proposed VLP system employs a  $4 \times 4$  imaging MIMO configuration that utilizes white LEDs as Txs and an imaging receiver with a PD array as Rx, in a similar arrangement to other MIMO-VLC imaging systems [27], [28].

#### 2.1. Transmitter

The VLP system utilizes four symmetrically distributed Txs, spaced 300 mm apart, serving as illumination sources and signal transmitters for position estimation. Each transmitter comprises an LED (Samsung LM561C) and a reflector (LEDiL EMILY-W), producing a 40° beam divergence. The LED operates with an average bias current of 75 mA, creating a luminous flux of 46 lumens. Modulating signals are generated by four arbitrary waveform generators (AWGs), whose outputs are converted into unipolar signals adding DC voltages using bias-Tees (MINI-CIRCUITS, ZFBT-4R2GW-FT+). Thus, a non-negative amplitude of the signal is ensured to modulate the intensity of each LED. Since the distance between the ceiling and the detector plane in the 2D VLP configuration is fixed at 1400 mm, the 40° divergence gives rise to a circular coverage area of approximately 817 mm<sup>2</sup> (510 mm radius) for each transmitter.

#### 2.2. Receiver

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Our imaging MIMO VLP configuration employs a plano-convex aspheric singlet lens (THORLABS, ACL2520U) with a 25 mm diameter and a 20.1 mm focal length as the imaging optics. A PD array with four independent elements is used as the receiver, as shown in Fig. 1. As stated in [29], the FoV can be improved by defocusing (i.e., placing receivers at the offset distance ( $f_{offset}$ ) towards the lens from the focal plane) instead of placing the receiver at the focal plane. Then, the receiver can achieve a wider FoV to support improved localization while maintaining a sufficient signal-to-noise ratio (SNR) for reliable positioning operation. Hence, we utilized the focal offset (foffset) of 4 mm, providing a FoV of 37.5°, which closely matches the transmitter beam divergence of 40°. The signal output from each PD is independently amplified by a trans-impedance amplifier (TIA) (MAX3665) followed by a low-noise amplifier (LNA) with a 20-dB gain (MINI-CIRCUITS, ZFL-1000LN+). A 4-channel digital oscilloscope captures the signal, with each channel corresponding to each amplified PD output, followed by offline processing. The maximum SNR in our setup is measured as 51.9 dB.

Table 2: Key experimental parameters for imaging VLP system.

	Parameter	Value	
LED	SA	MSUNG LM561C	
	Bias current $I_b$	75 mA	
	Bias voltage $V_{DC}$	3.3 V	
	Flux	43 lm @75 mA	
Reflector	LEDiL	CA11934_EMILY-W	
	External diameter	Ø 26 mm	

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	Parameter	Value			
	FWHM	40°			
RX lens	Thorlabs ACL2520U-A				
	Diameter	Ø25 mm			
	Focal Length $f_c$	20.1 mm			
	Back Focal Length $f_b$	12mm			
PD	First Sensor QP5.8-6-TO5				
	Number of elements	4			
	Active area of each PD	1.44 mm <sup>2</sup>			
	Responsivity	0.4 A/W @632 nm			
	Element gap	50 μm			
Amplifier	Mini-Circuits ZFL-1000LN+				
	Gain	19.9 dB			
	Noise Figure	2.9 dB			
	General				
	Discrete frequencies	200, 400, 600, 800 kHz			
	No. of transmitters, $M_{Tx}$	4			
	No. of receivers, $N_{Rx}$	4			
	Test Area	3D $540 \times 540 \times 120 \text{ mm}$			
		2D $410 \times 410 \text{ m}^2$			

#### 2.3. VLP channel

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For the MIMO configuration with  $N_{Tx}$  transmitters 63 and  $M_{Rx}$  receivers, the received signal can be calculated as: 64

$$S = HX + n; (1)$$

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where **H** is the  $N_{Tx} \times M_{Rx}$  channel matrix; **X** is the  $N_{Tx} \times 1$  transmitted signal vector; **n** is the  $M_{Rx} \times 1$  noise vector and **S** is the  $M_{Rx} \times 1$  received signal vector. Note that for the imaging optical MIMO communication system,  $M_{Rx} \geq N_{Tx}$ . However, such a requirement is not necessary for the VLP-only applic\_ations, though a higher number of Rx improves the positioning accuracy, as detailed in Section III. The channel gain from each Tx to each Rx, known as the channel **H**-matrix, is given by:

$$\mathbf{H} = \begin{bmatrix} h_{11} & h_{12} & \cdots & h_{1N_{Tx}} \\ h_{21} & h_{22} & \cdots & h_{2N_{Tx}} \\ \vdots & \vdots & \ddots & \vdots \\ h_{M_{Rx}1} & h_{M_{Rx}2} & \cdots & h_{M_{Rx}N_{Tx}} \end{bmatrix};$$
(2) 67

where  $h_{ij}$  is the channel gain from the  $j^{th}$  transmitter to the  $i^{th}$  receiver element. The channel gain information is related to RSS and can be used for positioning estimation.

For simplicity, we have employed frequency division multiplexing (FDM) with four discrete frequencies to distinguish the signals from individual LEDs instead of time division multiplexing (TDM), where a low-frequency sinusoid signal is transmitted in time sequence from each transmitter [27]. The Fast Fourier Transform (FFT) is applied to the received signal to compute the RSS corresponding to each transmitter. This operation is necessary to separate the FDM signals from multiple transmitters and prepare the inputs to the ANN for further processing.

#### 2.4. Artificial Neural Network model

As shown in Fig. 2, a fully connected feedforward 70 backpropagation supervised multi-layer perceptron (MLP) 71 ANN with one input layer, one hidden layer, and one output 72 layer is implemented for 2D/3D positioning estimation. The 73 number of neurons in the input layer equals 74  $M_{Rx} \times N_{Tx}$  corresponding to the channel **H**-matrix for a

particular position in (2). The output layer has a linear transfer function with two/three neurons corresponding to 2D/3D positioning, respectively. A detailed description of the ANN structure and corresponding training algorithm, including the optimization process for the hidden layer, can be found in [21]. Based on the optimization, the hidden layer has 32 neurons with a sigmoid transfer function. The sigmoid transfer function is selected for its capacity to introduce nonlinearities into the model, enabling the network to learn from the training data. A dataset containing different RSS matrices paired with their corresponding positions is used to train the ANN, which jointly represents the spatial distribution of RSS. During training, the network adjusts its weights to minimize errors between the estimated and actual positions. This is achieved by backpropagation, where the error gradient is propagated backward through the network, allowing the optimization of the neural network parameters [30]. While various algorithms can be used with backpropagation to update the weights and biases of the MLP-ANN, we specifically use the Levenberg-Marquardt algorithm due to its superior convergence speed and minimal epoch requirements compared to alternative methods [31].

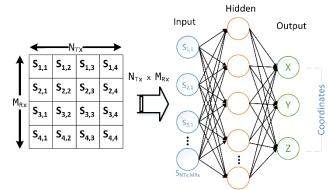
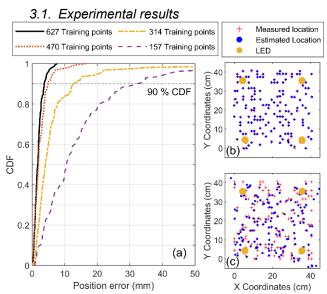


Fig. 2. Schematic diagram of ANN model for VLP.

#### 3. Results and Discussion



**Fig. 3.** (a) CDF of geometrical error of the proposed 2D VLP system using different numbers of training points. Location of the measured and estimated points by the ANN algorithm using different numbers of points: (b) 627 and (c) 157.

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The experiment aimed to evaluate the accuracy of the 46 proposed 2D/3D positioning system by conducting 47 measurements within the predefined areas. A grid of 2D 48 points with a geometric spacing of 15 mm was created to 49 evaluate the positioning accuracy. Hence, by measurements, 50 we collected 784 points within a  $410 \times 410 \text{ mm}^2$  area. This 51 geometric spacing was chosen to balance a detailed spatial 52 analysis with the practical constraints of measurement time. 53 The system was evaluated based on the geometrical error's 54 cumulative distribution function (CDF). Unless otherwise 55 specified, we will use a CDF of 0.9 to specify the positioning 56 error throughout the paper.

Fig. 3(a) shows the empirical CDF of the positioning 58 error for different numbers of ANN training data. Training 59 ANN with 80% of the dataset (627 points) resulted in a 60 positioning error of 3.7 mm. Conversely, reducing the 61 number of training data to 60%, 40%, and 20% of the dataset 62 led to performance degradations and increased positioning 63 errors of 4.9 mm, 12.6 mm, and 31.6 mm, respectively. 64 Figs. 3(b) and (c) present the spatial distributions of errors 65 across the test area. The figures compare the actual locations 66 (denoted by crosses) with the locations estimated by the 67 proposed system (represented by points). In Fig. 3(b), 627 68 training points were employed, and a smaller spread of the 69 estimated positions around the actual positions is observed, 70 indicating a higher overall positioning accuracy. In comparison, Fig. 3(c) shows the results when 157 training points were used, representing 20% of the dataset. This figure shows a sparser distribution of estimated positions, reflecting the degradation in system performance due to the smaller training dataset size.

To assess the 3D positioning accuracy, measurements were conducted within a  $540 \times 540 \times 120$  mm³ space. The grid is structured by points spaced 30 mm apart, with 361 points for each 2D plane and five different vertical levels, i.e., five 2D planes spaced 30 mm apart. The dataset comprises 1805 measurement points, 80% of which were used to train the ANN, and the remaining 20% were used to test and  $\frac{71}{72}$  validate the algorithm.

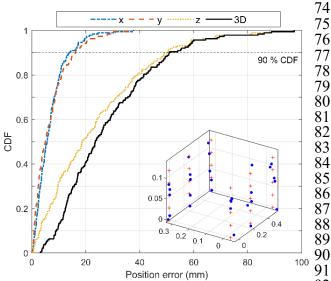


Fig. 4. CDF of geometrical error of the proposed 3D VLP system, 92 including the detail of individual x-, y-, and z- plane in the 3D 93 positioning. The inset shows the 3D scatter plot of the test and 94 estimated positioning points.

Fig. 4 shows the geometric error in the x-, y- and z-planes. The positioning accuracies in the x- and y-planes are comparable, where the error is close to 15 mm. As expected, the accuracy in x- and y-planes is similar to the value in 2D positioning experiments presented in Fig. 3 for a similar number of training points (314). In contrast, the z-plane has a significantly higher error of  $\sim$ 51 mm. This difference in error between the dimensions is due to the reduced number of training points in the z-plane, which results in lower positioning accuracy. The inset in Fig. 4 visually compares 30 test points (crosses) and the estimated positions (dots) in that plane.

Fig. 5 demonstrates the impact of the receiver diversity on the accuracy of 2D and 3D positioning. In the case of 2D positioning, the positioning errors are 30 mm, 6 mm, 5 mm, and 3.7 mm for one, two, three, and four PDs, respectively (see Fig. 5(a)). Similarly, Fig. 5(b) shows that the 3D positioning errors are 78 mm, 61 mm, 60 mm, and 51 mm for one, two, three, and four PDs, respectively. This clearly illustrates the advantage of a PD array receiver system with imaging optics in enhancing positioning accuracy. We anticipate further enhancements in 3D positioning by increasing the number of training points along the *z*-plane, as shown in the following section.

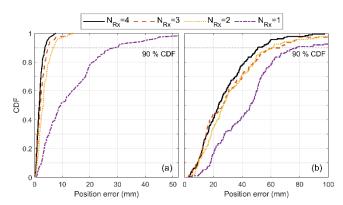


Fig. 5. CDF of geometrical error of the proposed VLP system with different diversity order  $N_{Rx}$  = 1-4 for: (a) 2D VLP and (b) 3D VLP.

#### 3.2. Simulation results

The ZEMAX OpticStudio software is employed to verify and further extend the experimental results. The simulation scenarios replicate the experimental configurations and components described in the previous section, while optical powers and shot noise are adjusted to match the laboratory measurement conditions. For 2D positioning, simulations were conducted in a  $410 \times 410 \text{ mm}^2$ area using the same grid and number of points as in the experiment. However, the imaging optics system was optimized by varying  $f_{offset}$ , i.e., distance of the lens from its focal point towards the detectors to modify the size of the image formed at the receiver, to study the impact on positioning accuracy.

Fig. 6(a) presents the CDF of the positioning error of the 2D-VLP system for  $f_{offset}$  ranging from 0 to 6 mm. As in the case of experimental work, 80% of the dataset is used to train the ANN, and the remaining 20% is used for testing. The solid line represents the CDF obtained from laboratory measurements, showing excellent agreement with simulation results, with a positioning error of 3.7 mm at  $f_{offset} = 4$  mm.

This correspondence validates the accuracy of the simulation 40 model in replicating the experimental setup, allowing further 41 extension of the experimental findings. The positioning error 42 initially decreases with increasing  $f_{offset}$ , reaching a minimum 43 error of approximately 2.6 mm at  $f_{offset} = 2$  mm; beyond this 44 point, the error begins to increase. For instance, the 45 positioning errors for  $f_{offset}$  of 3 mm and 4 mm are ~2.9 mm 46 and 3.5 mm, respectively.

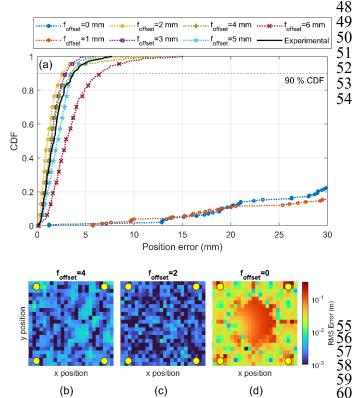
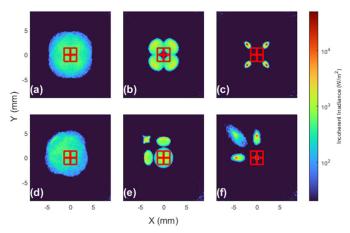


Fig. 6. (a) CDF of geometrical error from simulations for different 61 focal offsets between the lens and the photodetectors (black curve 62 corresponds to experimental results). Spatial distribution of the 63 geometrical error for: (b)  $f_{offset} = 4$  mm (c)  $f_{offset} = 2$  mm and (d) 64  $f_{offset} = 0$  mm. The yellow circles represent the location of the four 65

Figs. 6(b), (c), and (d) show the spatial distributions of  $_{68}^{67}$  the geometric error for an offset of 4 mm, 2 mm, and 0 mm,  $_{69}^{69}$  respectively. Figs. 6(b) and (c) reveal a relatively uniform  $_{70}^{70}$  distribution of positioning errors across the measured area. In  $_{71}^{70}$  contrast, Fig. 6(d) corresponding to  $_{foffset}^{60} = 0$  mm shows a  $_{72}^{70}$  notable variation in error distribution, with certain regions  $_{73}^{70}$  exhibiting significantly higher errors, particularly in the  $_{74}^{70}$  center. These areas of increased error are due to lower  $_{75}^{70}$  received signal intensity, as the image of Tx does not fall into  $_{76}^{70}$  any of the PDs due to narrow FoV (discussed further below).

Fig. 7 provides a further analysis of the imaging  $\frac{7}{8}$  MIMO VLP system. The simulation of the spatial intensity  $\frac{7}{9}$  distribution (incoherent irradiance W/m²) depicts the 80 received optical intensity at the image plane. The red lines 81 represent the PD array in the image plane. The first row (Figs. 82 7(a), (b), (c)) depicts the images formed when the detector is 83 located at the geometrical center formed by the transmitters. 84 The second row (Figs. 7(d), (e), (f)) represents the images 85 formed when the receiver is positioned directly below one of 86 the transmitters. Each column represents a  $f_{offset}$  of 4 mm, 87 2 mm and 0 mm (from left to right). As observed, the focal 88 offset significantly impacts the spatial distribution of 89

intensity. The clearest image is formed when the receiver plane is at the focal point ( $f_{offset} = 0$  mm). However, the images from the transmitters are outside the PDs, significantly reducing the received power. On the other hand, at  $f_{offset} = 4$  mm, the images formed from the transmitters overlap significantly and are difficult to distinguish (this overlap leads to substantial inter-channel interference, resulting in a higher condition number for the MIMO H-matrix). In contrast,  $f_{offset} = 2$  mm proves to be the optimal configuration, showing a clearly separated image from the four transmitters. Hence, as in the case of the imaging MIMO-VLC system [31], the condition number of channel H-matrix affected the positioning accuracy, and optimization of the optic system is necessary to obtain the best condition number.



**Fig. 7.** Simulation of the image formed by the transmitters at the detector plane for different  $f_{offset}$  and different locations. First row: The receiver is at the geometric center of the four transmitters and  $f_{offset}$  is: (a) 4 mm, (b) 2 mm, and (c) 0 mm. Second row: The receiver is located directly under one of the transmitters and  $f_{offset}$  is: (d) 4 mm, (e) 2 mm, and (f) 0 mm. The red lines represent the PD array of four elements at the imaging plane.

Furthermore, the impact of SNR on the VLP performance is shown in Fig. 8, where the position accuracy has been estimated for the best-case configuration  $f_{offset} = 2$  mm under an SNR range of 25–60 dB. As expected, the accuracy of the system decreases as the SNR decreases. In particular, the accuracy decreases from ~2 mm to ~16.9 mm when the maximum SNR reduces from 60.9 dB to 25.9 dB. However, the SNR has only a marginal impact on the positioning at high SNR, e.g., position accuracy decreases from 2 mm to 2.4 mm when SNRs are 60.9 dB and 50.9 dB, respectively. The insets illustrate the SNR distribution over the measurement area for a single transmitter and a single PD from the array. Note the significant drops in SNRs towards the edge of coverage areas.

Finally, simulations for both 2D and 3D positioning were carried out in an enlarged area of  $600 \times 600 \times 240 \text{ mm}^3$  under the same configuration as employed in the laboratory ( $f_{offset} = 4 \text{ mm}$ ) for estimating the potential accuracy of the experimental 3D positioning system with a larger number of dataset points. The simulation grid is structured by points spaced 15 mm apart in the horizontal and vertical dimensions. This configuration results in a total of 1681 points in each of the 2D planes. The grid was divided vertically into 17 different levels ( $\pm 120 \text{ mm}$  from the 2D level); thus, the simulation dataset comprises 28577 points for 3D. As in all other cases, 80 % of the dataset (22861) is used to train the

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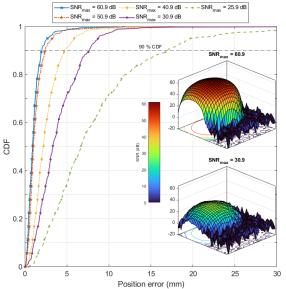
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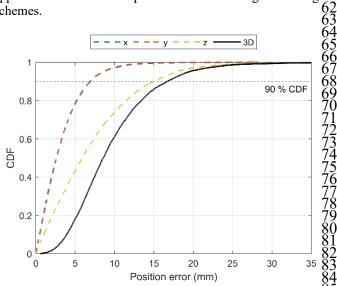
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ANN, and the remaining 20% is used for testing (5715). 27 Fig. 9 displays the CDF of the 3D positioning along the x-, y- 28 and z-planes, as well as the combined total positioning error. 29



**Fig. 8.** CDF of the geometric error from simulations with a 50 configuration of  $f_{offset} = 2$  mm and 1513 dataset points for different 51 SNR = 60.9–25.9 dB. Insets show the SNR distribution within the measurement area for one LED and one PD from the array for the 52 best-case scenario of: (i) 60.9 dBm and (ii) 30.9 dBm.

The x- and y-planes errors are similar and relatively 54 low, with positioning errors of ~6.9 mm. The z-plane, on the 55 other hand, exhibits a greater error of 15.0 mm. The total 3D 57 positioning error, which combines the errors of all three axes, 58 is 16.8 mm, which is significantly lower than the 59 experimental results of 51 mm. This improvement is due to 59 the higher density of training points in the z-plane in the simulation, which allows for a more accurate representation of the measurement space and shows the potential of our approach for further improvement under larger training 61 schemes.



**Fig. 9** CDF of geometrical error for 3D VLP from simulations with a configuration of  $f_{offset} = 4$  mm and 28577 dataset points.

#### 4. Conclusions

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This paper presents an experimental demonstration of a high-accuracy 2D and 3D VLP system using an optical imaging MIMO system with supervised ANN. Experimental results show the impact of the number of training points and the spatial diversity on positioning accuracy, leading to 3.7 mm and 51 mm for 2D and 3D positioning accuracy, respectively.

The experimental work is validated and further extended by simulations. The simulation results allow to estimate the required number of training points and evaluate the impact of the system noise. Furthermore, the study underscores the critical role of optimizing the imaging optics, particularly the focal offset, in achieving high-accuracy positioning. Using the experimental parameters under an system configuration, the optimized demonstrated a minimum positioning error of 2.0 mm and 9.3 mm for 2D and 3D positioning, respectively. Therefore, highaccuracy 2D/3D positioning has been demonstrated in a scalable imaging MIMO configuration as a promising solution for integrated positioning and communications applications. These advantages lead the proposed VLP imaging MIMO system as a promising solution for future 6G networks requiring high-precision indoor positioning capabilities.

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